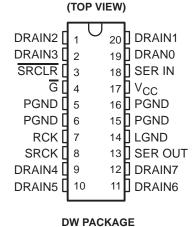
- Low r_{DS(on)} . . . 1 Ω Typ
- Output Short-Circuit Protection
- Avalanche Energy . . . 75 mJ
- Eight 350-mA DMOS Outputs
- 50-V Switching Capability
- Devices Are Cascadable
- Low Power Consumption

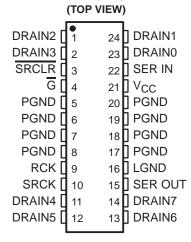
description

The TPIC6A595 is a monolithic, high-voltage, high-current power logic 8-bit shift register designed for use in systems that require relatively high load power. The device contains a built-in voltage clamp on the outputs for inductive transient protection. Power driver applications include relays, solenoids, and other medium-current or high-voltage loads. Each open-drain DMOS transistor features an independent chopping current-limiting circuit to prevent damage in the case of a short circuit.

This device contains an 8-bit serial-in, parallel-out shift register that feeds an 8-bit, D-type storage register. Data transfers through both the shift and storage registers on the rising edge of the shift-register clock (SRCK) and the register clock (RCK), respectively. The storage register transfers data to the output buffer when shift-register clear (SRCLR) is high. When SRCLR is low, the input shift register is cleared. When output



NE PACKAGE



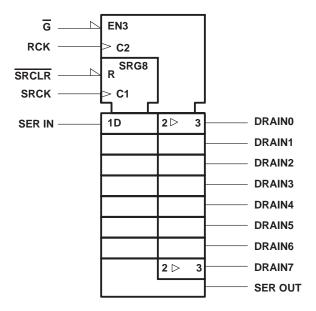
enable (\overline{G}) is held high, all data in the output buffers is held low and all drain outputs are off. When \overline{G} is held low, data from the storage register is transparent to the output buffers. The serial output (SER OUT) allows for cascading of the data from the shift register to additional devices.

Outputs are low-side, open-drain DMOS transistors with output ratings of 50 V and a 350-mA continuous sink current capability. When data in the output buffers is low, the DMOS-transistor outputs are off. When data is high, the DMOS-transistor outputs have sink current capability.

Separate power ground (PGND) and logic ground (LGND) terminals are provided to facilitate maximum system flexibility. All PGND terminals are internally connected, and each PGND terminal must be externally connected to the power system ground in order to minimize parasitic impedance. A single-point connection between LGND and PGND must be made externally in a manner that reduces crosstalk between the logic and load circuits.

The TPIC6A595 is offered in a thermally-enhanced dual-in-line (NE) package and a wide-body surface-mount (DW) package. The TPIC6A595 is characterized for operation over the operating case temperature range of -40° C to 125° C.

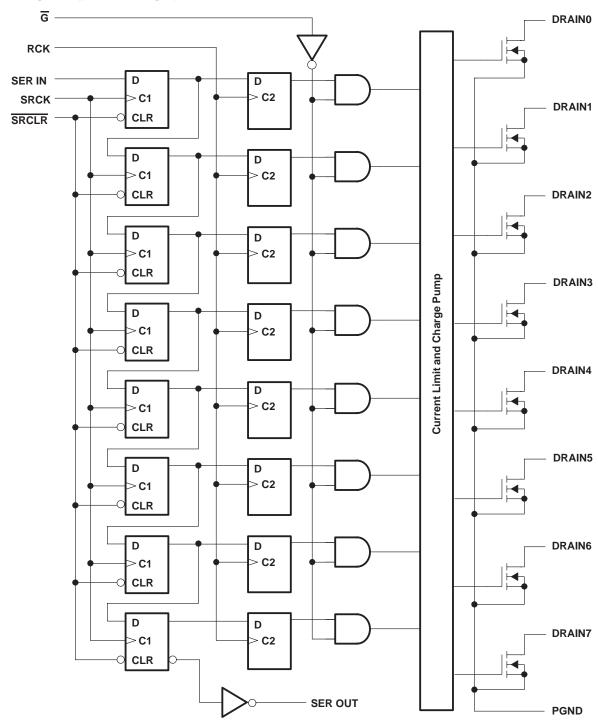
logic symbol†



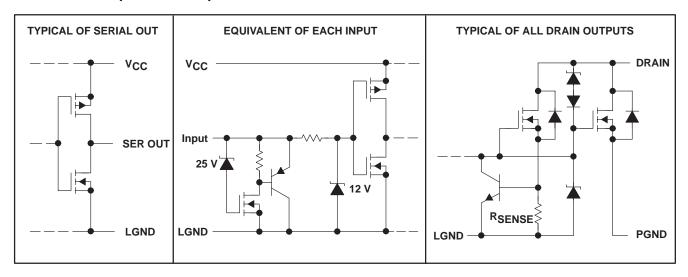
[†] This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.



logic diagram (positive logic)



schematic of inputs and outputs



absolute maximum ratings over recommended operating case temperature range (unless otherwise noted)†

Logic supply voltage, V _{CC} (see Note 1)	7 V
Logic input voltage range, V _I	0.3 V to 7 V
Power DMOS drain-to-source voltage, V _{DS} (see Note 2)	50 V
Continuous source-drain diode anode current	1 A
Pulsed source-drain diode anode current (see Note 3)	2 A
Pulsed drain current, each output, all outputs on, IDn, TA = 25°C (see Note 3	3) 1.1 A
Continuous drain current, each output, all outputs on, I _{Dn.} T _A = 25°C	350 mA
Peak drain current, single output, T _A = 25°C (see Note 3)	1.1 A
Single-pulse avalanche energy, EAS (see Figure 6)	75 mJ
Avalanche current, I _{AS} (see Note 4)	600 mA
Continuous total dissipation	See Dissipation Rating Table
Operating case temperature range, T _C	–40°C to 125°C
Operating virtual junction temperature range, T _J	–40°C to 150°C
Storage temperature range, T _{stg}	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values are with respect to LGND and PGND.
 - 2. Each power DMOS source is internally connected to PGND.
 - 3. Pulse duration \leq 100 μ s and duty cycle \leq 2 %.
 - 4. DRAIN supply voltage = 15 V, starting junction temperature $(T_{JS}) = 25$ °C, L = 210 mH, $I_{AS} = 600$ mA (see Figure 6).

DISSIPATION RATING TABLE

PACKAGE	$T_C \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _C = 25°C	T _C = 125°C POWER RATING
DW	1750 mW	14 mW/°C	350 mW
NE	2500 mW	20 mW/°C	500 mW



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recommended operating conditions

	MIN	MAX	UNIT
Logic supply voltage, V _{CC}	4.5	5.5	V
High-level input voltage, V _{IH}	0.85 V _C C	VCC	V
Low-level input voltage, V _{IL}	0	0.15 V _{CC}	V
Pulsed drain output current, T _C = 25°C, V _{CC} = 5 V (see Notes 3 and 5)	-1.8	0.6	Α
Setup time, SER IN high before SRCK↑, t _{SU} (see Figure 2)	10		ns
Hold time, SER IN high after SRCK↑, th (see Figure 2)	10		ns
Pulse duration, t _W (see Figure 2)	20		ns
Operating case temperature, T _C	-40	125	°C

electrical characteristics, V_{CC} = 5 V, T_{C} = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{(BR)DSX}	Drain-to-source breakdown voltage	I _D = 1 mA	50			V
V _{SD}	Source-to-drain diode forward voltage	I _F = 350 mA, See Note 3		0.8	1.1	V
Vall	High-level output voltage,	$I_{OH} = -20 \mu\text{A}$	V _{CC} -0.1	Vcc		٧
Vон	SER OUT	$I_{OH} = -4 \text{ mA}$	V _{CC} -0.5	V _{CC} -0.2		
\/a:	Low-level output voltage,	I _{OL} = 20 μA		0	0.1	V
VOL	SER OUT	I _{OL} = 4 mA		0.2	0.5	٧
lн	High-level input current	$V_I = V_{CC}$			1	μΑ
I _{IL}	Low-level input current	V _I = 0			-1	μΑ
IO(chop)	Output current at which chopping starts	T _C = 25°C, See Note 5 and Figures 3 and 4	0.6	0.8	1.1	Α
Icc	Logic supply current	$I_O = 0$, $V_I = V_{CC}$ or 0		0.5	5	mA
ICC(FRQ)	Logic supply current at frequency	$f_{SRCK} = 5 \text{ MHz},$ $I_O = 0,$ $C_L = 30 \text{ pF},$ $V_I = V_{CC} \text{ or } 0,$ $V_{CC} = 5 \text{ V},$ See Figure 7		1.3		mA
I _(nom)	Nominal current	$V_{DS(on)} = 0.5 \text{ V},$ $I_{(nom)} = I_D, T_C = 85^{\circ}C,$ $V_{CC} = 5 \text{ V},$ See Notes 5, 6, and 7		350		mA
	Drain gurrent off state	$V_{DS} = 40 \text{ V}, T_{C} = 25^{\circ}\text{C}$		0.1	1	
ΙD	Drain current, off-state	$V_{DS} = 40 \text{ V}, T_{C} = 125^{\circ}\text{C}$		0.2	5	μΑ
	Static drain-source on-state resistance	$I_D = 350 \text{ mA}, T_C = 25^{\circ}\text{C}$		1	1.5	
r _{DS(on)}		I _D = 350 mA, T _C = 125°C See Notes 5 and 6 and Figures 10 and 11		1.7	2.5	Ω
		$I_D = 350 \text{ mA}, T_C = 40^{\circ}\text{C}$				

NOTES: 3. Pulse duration \leq 100 μ s and duty cycle \leq 2%.

- 5. Technique should limit $T_J T_C$ to 10°C maximum.
- 6. These parameters are measured with voltage-sensing contacts separate from the current-carrying contacts.
- 7. Nominal current is defined for a consistent comparison between devices from different sources. It is the current that produces a voltage drop of 0.5 V at T_C = 85°C.



TPIC6A595 **POWER LOGIC 8-BIT SHIFT REGISTER**

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switching characteristics, V_{CC} = 5 V, T_{C} = 25°C

PARAMETER		TEST CONDITIONS	MIN TY	P MAX	UNIT
tPHL	Propagation delay time, high-to-low-level output from \overline{G}	C _L = 30 pF, I _D = 350 mA, See Figures 1, 2, and 12	3	0	ns
tPLH	Propagation delay time, low-to-high-level output from \overline{G}		12	5	ns
t _r	Rise time, drain output		6	0	ns
tf	Fall time, drain output		3	0	ns
ta	Reverse-recovery-current rise time	$I_F = 350$ mA, di/dt = 20 A/ μ s, See Notes 5 and 6 and Figure 5	10	0	ns
t _{rr}	Reverse-recovery time		30	0	ns

thermal resistance

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT		
R _{θJC} Thermal resistance, junction-to-case	DW	DW	All eight outputs with equal power		10	°C/W	
	NE	All eight outputs with equal power		10	C/VV		
R _{0JA} Thermal resistance, junction-to-ambient	Lain Thermal resistance, junction-to-ambient	DW	All eight outputs with equal power		50	°C/W	
		NE	All eight outputs with equal power		50	C/VV	



NOTES: 5. Technique should limit T_J – T_C to 10°C maximum.

6. These parameters are measured with voltage-sensing contacts separate from the current-carrying contacts.

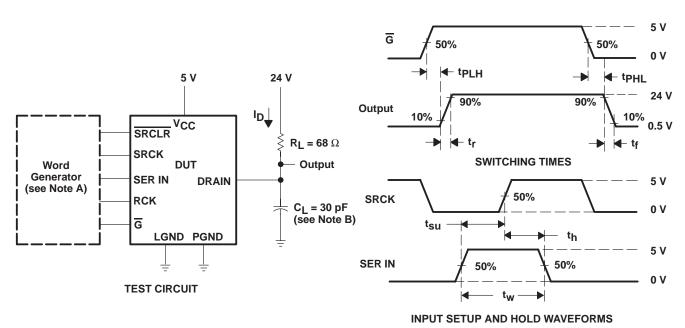
VOLTAGE WAVEFORMS

PARAMETER MEASUREMENT INFORMATION 5 V 24 V **SRCK** 0 V ID↓ VCC 5 V SRCLR G $R_L = 68 \Omega$ 0 V SRCK DUT 5 V Output **SER IN** Word 0 V **DRAIN** Generator **SER IN** 5 V (see Note A) **RCK** $C_L = 30 pF$ 0 V **RCK** (see Note B) G SRCLR 0 V LGND PGND 24 V **DRAIN 0, 1, 4, 5** 0.5 V 24 V **TEST CIRCUIT DRAIN 2, 3, 6, 7**

NOTES: A. The word generator has the following characteristics: $t_{\Gamma} \le 10$ ns, $t_{W} = 300$ ns, pulsed repetition rate (PRR) = 5 kHz, $Z_{O} = 50~\Omega$.

B. CL includes probe and jig capacitance.

Figure 1. Resistive Load Operation



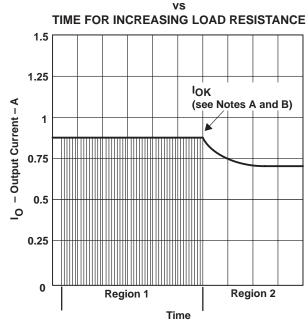
NOTES: A. The word generator has the following characteristics: $t_{\Gamma} \le 10$ ns, $t_{\tilde{\Gamma}} \le 10$ ns, $t_{W} = 300$ ns, pulsed repetition rate (PRR) = 5 kHz, $Z_{\tilde{Q}} = 50~\Omega$.

B. C_L includes probe and jig capacitance.

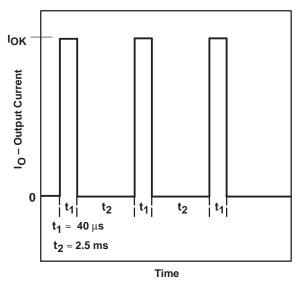
Figure 2. Test Circuit, Switching Times, and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

OUTPUT CURRENT



REGION 1 CURRENT WAVEFORM



First output current pulses after turn-on in chopping mode with resistive load.

- NOTES: A. Figure 3 illustrates the output current characteristics of the device energizing a load having initially low, increasing resistance, e.g., an incandescent lamp. In region 1, chopping occurs and the peak current is limited to I_{OK}. In region 2, output current is continuous. The same characteristics occur in reverse order when the device energizes a load having an initially high, decreasing resistance.
 - B. Region 1 duty cycle is approximately 2%.

Figure 3. Chopping-Mode Characteristics

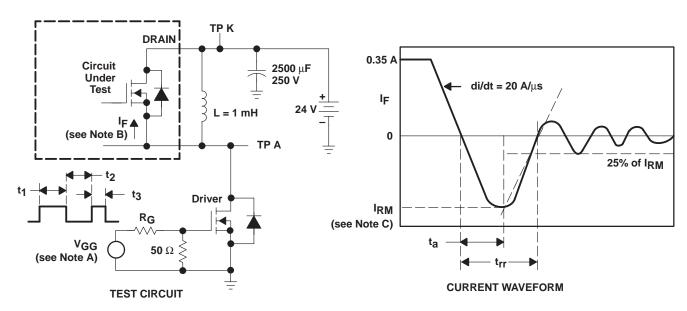
OUTPUT CURRENT LIMIT

CASE TEMPERATURE 1.5 V_{CC} = 5.5 V 1.2 Output Current Limit – A 0.9 $V_{CC} = 4.5$ 0.6 _0 0.3 - 50 - 25 25 50 75 100 125 150 T_C - Case Temperature - °C

Figure 4

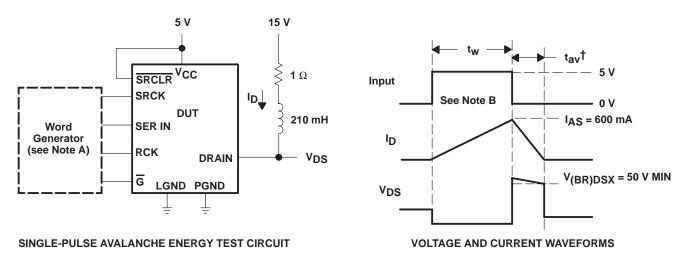


PARAMETER MEASUREMENT INFORMATION



- NOTES: A. The VGG amplitude and RG are adjusted for di/dt = 20 A/ μ s. A VGG double-pulse train is used to set IF = 0.35 A, where t₁ = 10 μ s, t₂ = 7 μ s, and t₃ = 3 μ s.
 - B. The DRAIN terminal under test is connected to the TP K test point. All other terminals are connected together and connected to the TP A test point.
 - C. I_{RM} = maximum recovery current

Figure 5. Reverse-Recovery-Current Test Circuit and Waveforms of Source-Drain Diode



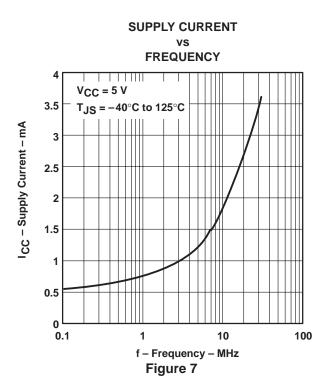
† Non JEDEC symbol for avalanche time.

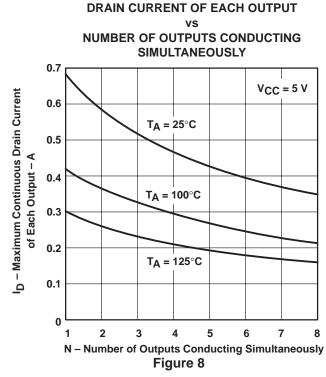
- NOTES: A. The word generator has the following characteristics: $t_f \le 10$ ns, $t_f \le 10$ ns, $Z_O = 50 \Omega$.
 - B. Input pulse duration, t_{W} , is increased until peak current $I_{AS} = 600$ mA. Energy test level is defined as $E_{AS} = (I_{AS} \times V_{(BR)DSX} \times t_{aV})/2 = 75$ mJ.

Figure 6. Single-Pulse Avalanche Energy Test Circuit and Waveforms



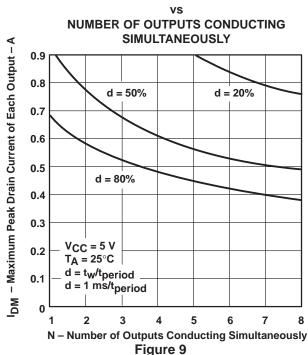
TYPICAL CHARACTERISTICS



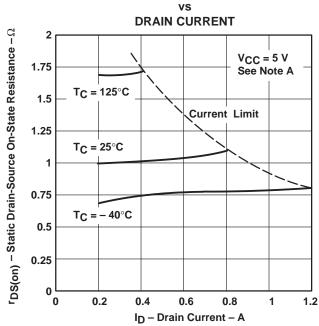


MAXIMUM CONTINUOUS

MAXIMUM PEAK DRAIN CURRENT OF EACH OUTPUT



STATIC DRAIN-SOURCE ON-STATE RESISTANCE

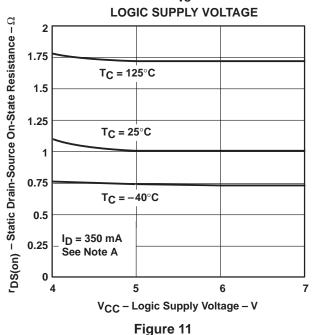


NOTE A: Technique should limit $T_J - T_C$ to 10°C maximum. Figure 10



TYPICAL CHARACTERISTICS

STATIC DRAIN-SOURCE ON-STATE RESISTANCE vs



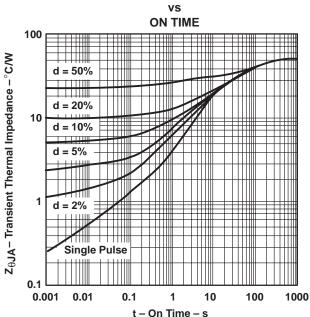
SWITCHING TIME CASE TEMPERATURE 140 $I_D = 350 \text{ mA}$ See Note A 120 ^tPLH Switching Time - ns 100 80 60 ^tPHL 40 tf 20 -5050 100 150 T_C - Case Temperature - °C

Figure 12

NOTE A: Technique should limit $T_J - T_C$ to 10°C maximum.

THERMAL INFORMATION

NE PACKAGE TRANSIENT THERMAL IMPEDANCE



The single-pulse curve represents measured data. The curves for various pulse durations are based on the following equation:

$$Z_{\theta JA} = \left| \frac{t_w}{t_c} \right| R_{\theta JA} + \left| 1 - \frac{t_w}{t_c} \right| Z_{\theta}(t_w + t_c)$$

Where:

 $\mathbf{Z}_{\theta}(\mathbf{t_{w}}) = \underset{\text{for } t = t_{W}}{\text{single-pulse thermal impedance}}$

 $+ Z_{\theta}(t_{\mathbf{W}}) - Z_{\theta}(t_{\mathbf{C}})$

 $Z_{\theta}(t_{\text{C}}) \ = \ \text{the single-pulse thermal impedance} \\ \text{for } t = \ t_{\text{C}} \ \text{seconds}$

 $Z_{\theta}(t_W + t_C) = \text{the single-pulse thermal impedance}$ for $t = t_W + t_C \text{ seconds}$

 $d = t_W/t_C$ $t_W \longrightarrow \begin{bmatrix} t_C & & \\ & & \end{bmatrix}$

Figure 13



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